

AN11514

BGU8H1 LTE LNA evaluation board

Rev. 3 — 22 January 2016

Application note

Document information

| Info | Content |
|----------------------------|---|
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| Abstract | This document explains the BGU8H1 LTE LNA evaluation board |
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| Contact information | For more information, please visit: http://www.nxp.com |



Revision history

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1. Introduction

NXP Semiconductors' BGU8H1 LTE LNA Evaluation Board is designed to evaluate the performance of the LTE LNA using:

- NXP Semiconductors' BGU8H1 LTE Low Noise Amplifier
- A matching inductor
- A decoupling capacitor

NXP Semiconductors' BGU8H1 is a low-noise amplifier for LTE receiver applications in a plastic, leadless 6 pin, extremely thin small outline SOT1232 at 1.1 x 0.7 x 0.37mm, 0.4mm pitch. The BGU8H1 features gain of 13 dB and a noise figure of 0.9 dB at a current consumption of 5 mA. Its superior linearity performance removes interference and noise from co-habitation cellular transmitters, while retaining sensitivity. The LNA components occupy a total area of approximately 4 mm².

In this document, the application diagram, board layout, bill of materials, and typical results are given, as well as some explanations on LTE related performance parameters like input third-order intercept point IIP3, gain compression and noise.

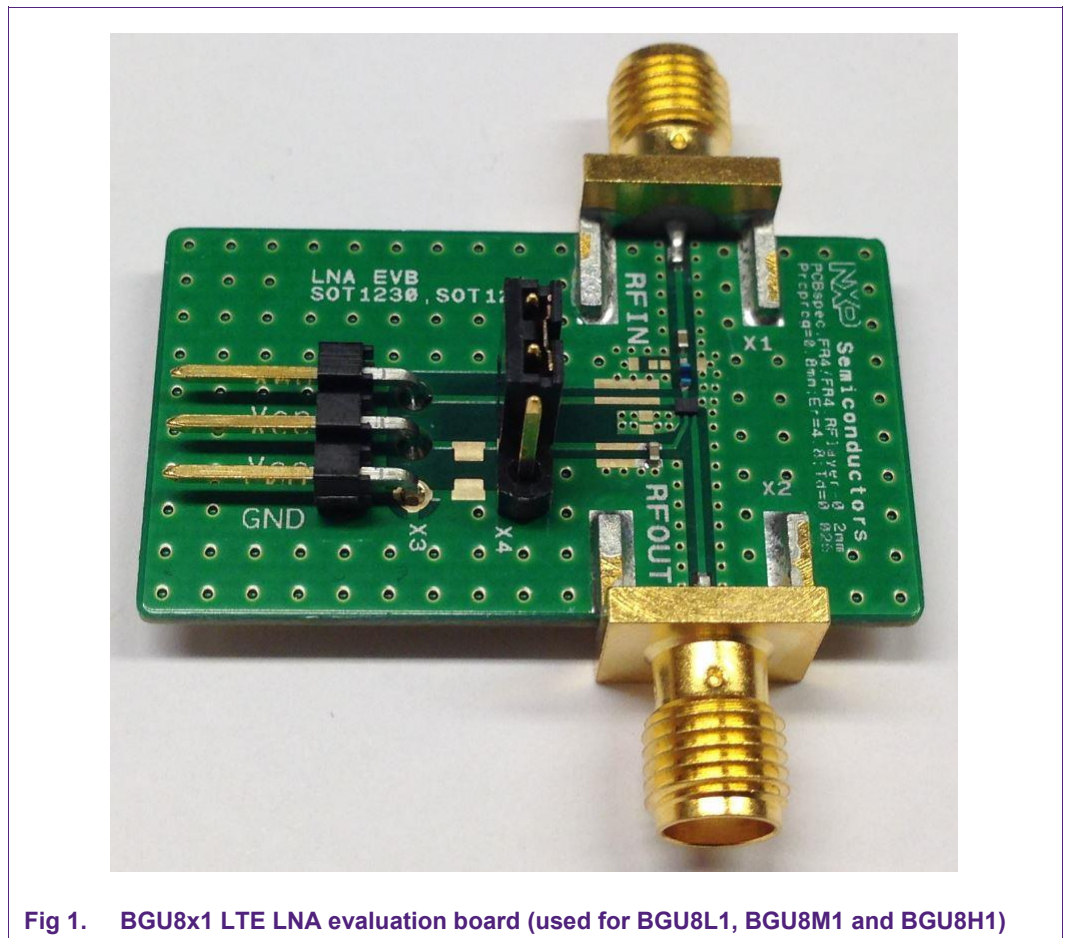


Fig 1. BGU8x1 LTE LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1)

2. General description

Modern cellular phones have multiple radio systems, so problems like co-habitation are quite common. Since the LTE diversity antenna needs to be placed far from the main antenna to ensure the efficiency of the channel, a low noise amplifier close to the antenna is used to compensate the track-losses (and SAW-filter losses when applicable) on the printed circuit board. A LTE receiver implemented in a mobile phone requires a low current consumption and low Noise Figure. All the different transmit signals that are active in smart phones and tablets can cause problems like inter-modulation and compression. Therefore also a high linearity is required.

3. BGU8H1 LTE LNA evaluation board

The BGU8H1LNA evaluation board simplifies the RF evaluation of the BGU8H1 LTE LNA applied in a LTE front-end, often used in mobile cell phones. The evaluation board enables testing of the device RF performance and requires no additional support circuitry. The board is fully assembled with the BGU8H1 including the input series inductor and decoupling capacitor. The board is supplied with two SMA connectors for input and output connection to RF test equipment. The BGU8H1 can operate from a 1.5 V to 3.1 V single supply and consumes typical 5 mA.

3.1 Application Circuit

The circuit diagram of the evaluation board is shown in Fig 2. With jumper JU1 the enable input can be connected either to Vcc or GND.

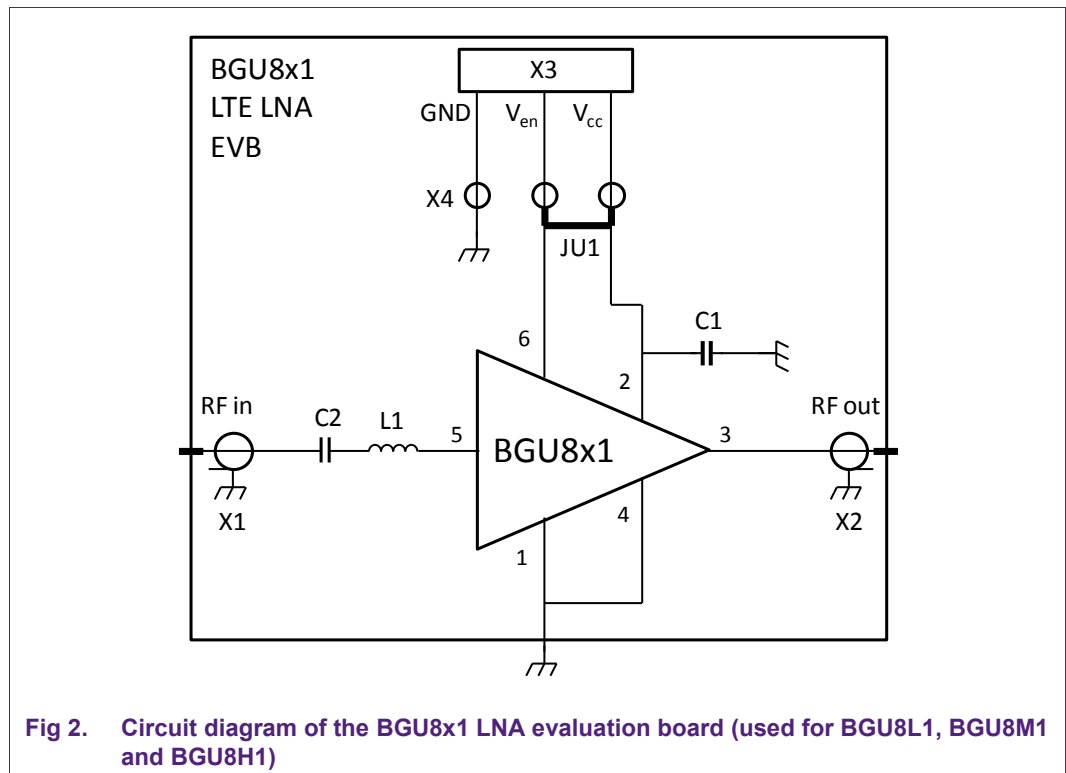


Fig 2. Circuit diagram of the BGU8x1 LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1)

3.2 PCB Layout

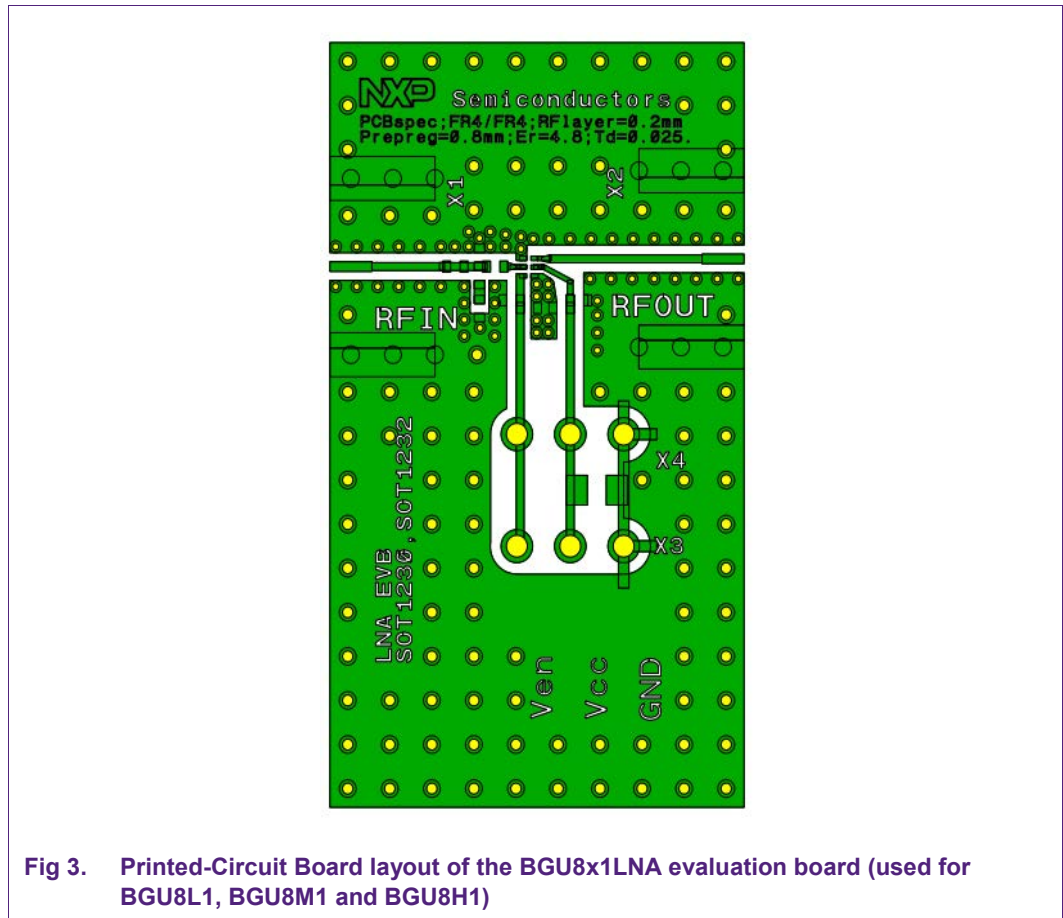


Fig 3. Printed-Circuit Board layout of the BGU8x1LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1)

A good PCB layout is an essential part of an RF circuit design. The LNA evaluation board of the BGU8H1 can serve as a guideline for laying out a board using the BGU8H1. Use controlled impedance lines for all high frequency inputs and outputs. Bypass Vcc with decoupling capacitors, preferably located as close as possible to the device. For long bias lines it may be necessary to add decoupling capacitors along the line further away from the device. Proper grounding of the GND pins is also essential for good RF performance. Either connect the GND pins directly to the ground plane or through vias, or do both, which is recommended. The material that has been used for the evaluation board is FR4 using the stack shown in Fig 4.

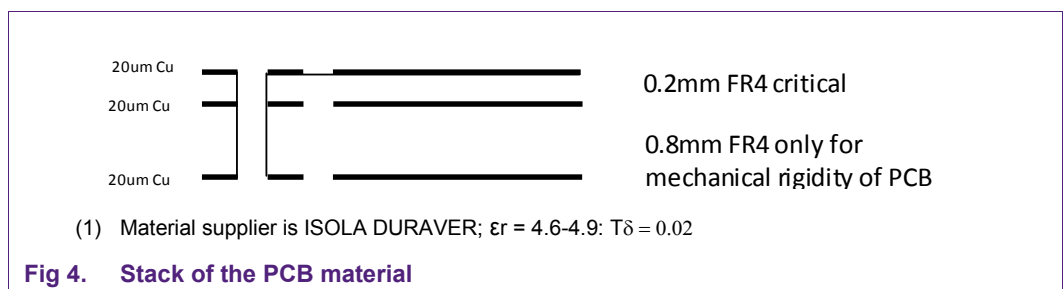


Fig 4. Stack of the PCB material

4. Bill of materials

Table 1. BOM of the BGU8H1 LTE LNA evaluation board

| Designator | Description | Footprint | Value | Supplier Name/type | Comment |
|------------|------------------|--|-------|---|--|
| F | BGU8H1 | 1.1 x 0.7 x 0.37mm ³ , 0.4mm pitch | | NXP | SOT1232 |
| PCB | | 20 x 35mm | | BGU8H1 LTE LNA EV Kit | |
| C1 | Capacitor | 0402 | 1nF | Murata GRM1555 | Decoupling |
| C2 | Capacitor | 0402 | 1nF | Murata GRM1555 | Decoupling |
| L1 | Inductor | 0402 | 3.3nH | Murata LQW15 | Input matching |
| X1, X2 | SMA RD connector | - | - | Johnson, End launch SMA 142-0701-841 | RF input/ RF output |
| X3 | DC header | - | - | Molex, PCB header, Right Angle, 1 row, 3 way 90121-0763 | Bias connector |
| X4 | JUMPER Stage | - | - | Molex, PCB header, Vertical, 1 row, 3 way 90120-0763 | Connect Ven to Vcc or separate Ven voltage |
| JU1 | JUMPER | | | | |

4.1 BGU8H1

NXP Semiconductors' BGU8H1 LTE low noise amplifier is designed for the LTE frequency band. The integrated biasing circuit is temperature stabilized, which keeps the current constant over temperature. It also enables the superior linearity performance of the BGU8H1. The BGU8H1 is also equipped with an enable function that allows it to be controlled via a logic signal. In disabled mode it consumes less than 1 μ A.

The output of the BGU8H1 is internally matched between 2300 MHz and 2690 MHz whereas only one series inductor at the input is needed to achieve the best RF performance. The output is AC coupled via an integrated capacitor.

It requires only two external components to build a LTE LNA having the following advantages:

- Low noise
- System optimized gain
- High linearity under jamming
- 1.1 x 0.7 x 0.37, 0.4mm pitch: SOT1232
- Low current consumption
- Short power settling time

4.2 Series inductor

The evaluation board is supplied with Murata LQW15 series inductor of 3.3 nH. This is a wire wound type of inductor with high quality factor (Q) and low series resistance (Rs). This type of inductor is recommended in order to achieve the best noise performance. High Q inductors from other suppliers can be used. If it is decided to use other low cost inductors with lower Q and higher Rs the noise performance will degrade.

5. Required Equipment

In order to measure the evaluation board the following is necessary:

- ✓ DC Power Supply op to 30 mA at 1.5 V to 3.1 V
- ✓ Two RF signal generators capable of generating RF signals at the LTE operating frequency between 2300 MHz and 2690 MHz.
- ✓ An RF spectrum analyzer that covers at least the operating frequency between 2300 MHz and 2690 MHz as well as a few of the harmonics. Up to 6 GHz should be sufficient.
“Optional” a version with the capability of measuring noise figure is convenient
- ✓ Amp meter to measure the supply current (optional)
- ✓ A network analyzer for measuring gain, return loss and reverse isolation
- ✓ Noise figure analyzer and noise source
- ✓ Directional coupler
- ✓ Proper RF cables

6. Connections and setup

The BGU8H1 LTE LNA evaluation board is fully assembled and tested (see Fig 5). Please follow the steps below for a step-by-step guide to operate the LNA evaluation board and testing the device functions.

1. Connect the DC power supply to the V_{cc} and GND terminals. Set the power supply to the desired supply voltage, between 1.5 V and 3.1 V, but never exceed 3.1 V as it might damage the BGU8H1.
2. Jumper JU1 is connected between the V_{cc} terminal of the evaluation board and the V_{en} pin of the BGU8H1.
3. Connect the RF signal generator and the spectrum analyzer to the RF input and the RF output of the evaluation board, respectively. Do not turn on the RF output of the signal generator yet, set it to approximately -40 dBm output power at center frequency of the wanted LTE-ban and set the spectrum analyzer at the same center frequency and a reference level of 0 dBm.
4. Turn on the DC power supply and it should read approximately 4..5 mA.
5. Enable the RF output of the generator: The spectrum analyzer displays a tone around -27 dBm.
6. Instead of using a signal generator and spectrum analyzer one can also use a network analyzer in order to measure gain as well as in- and output return loss, P1dB and IP3 (see Fig 6).
7. For noise figure evaluation, either a noise figure analyzer or a spectrum analyzer with noise option can be used. The use of a 5 dB noise source, like the Agilent 364B is recommended. When measuring the noise figure of the evaluation board, any kind of adaptors, cables etc between the noise source and the evaluation board should be minimized, since this affects the noise figure (see Fig 7).

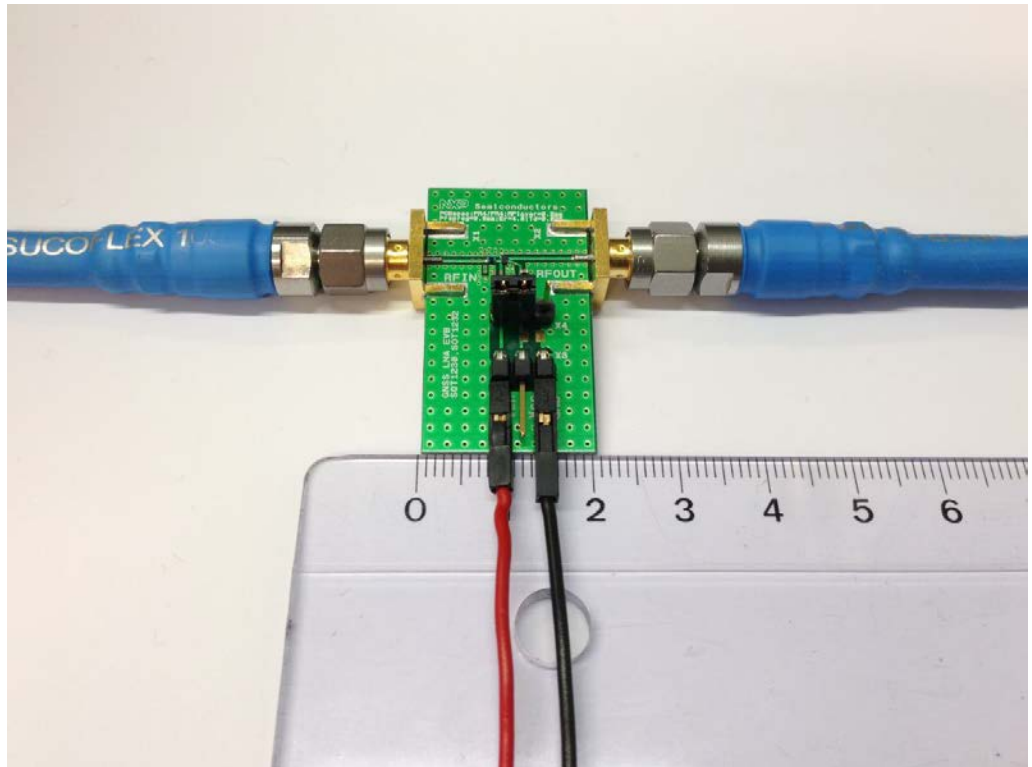


Fig 5. Evaluation board including its connections

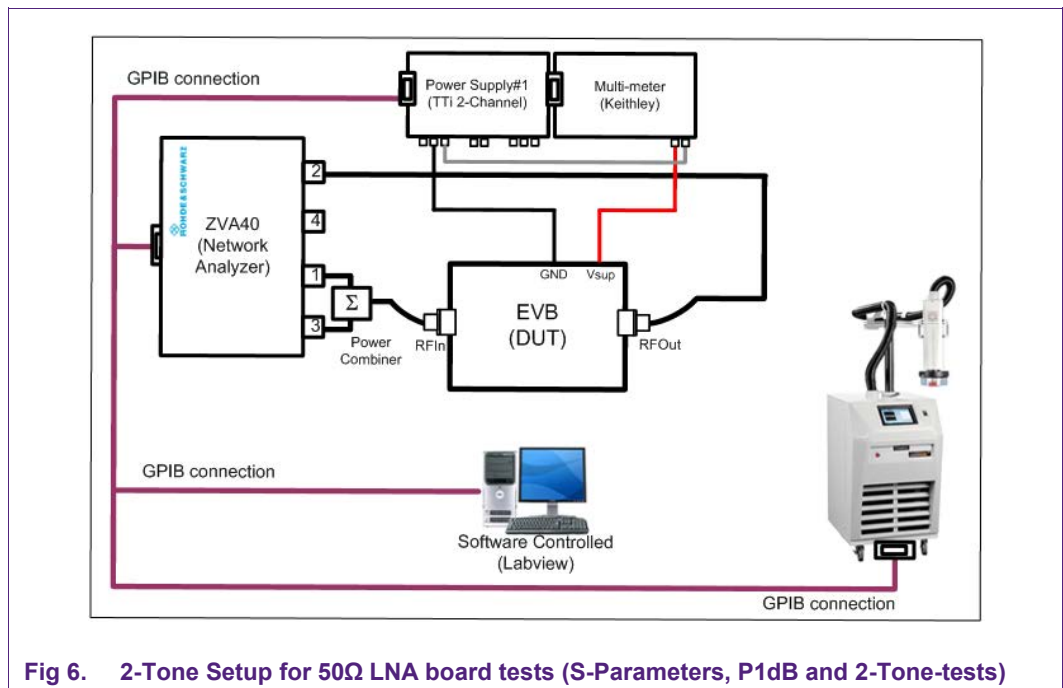


Fig 6. 2-Tone Setup for 50Ω LNA board tests (S-Parameters, P1dB and 2-Tone-tests)

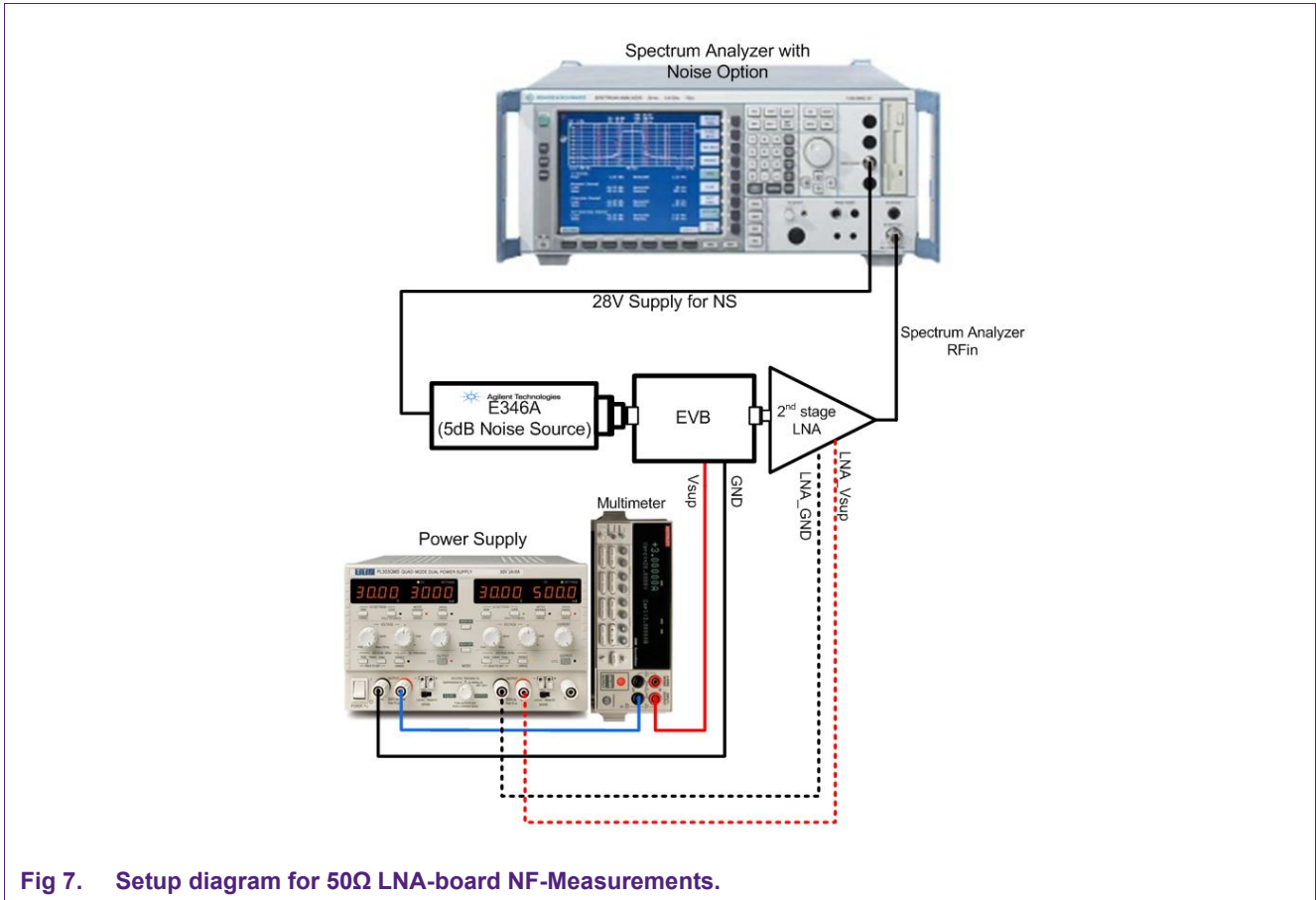
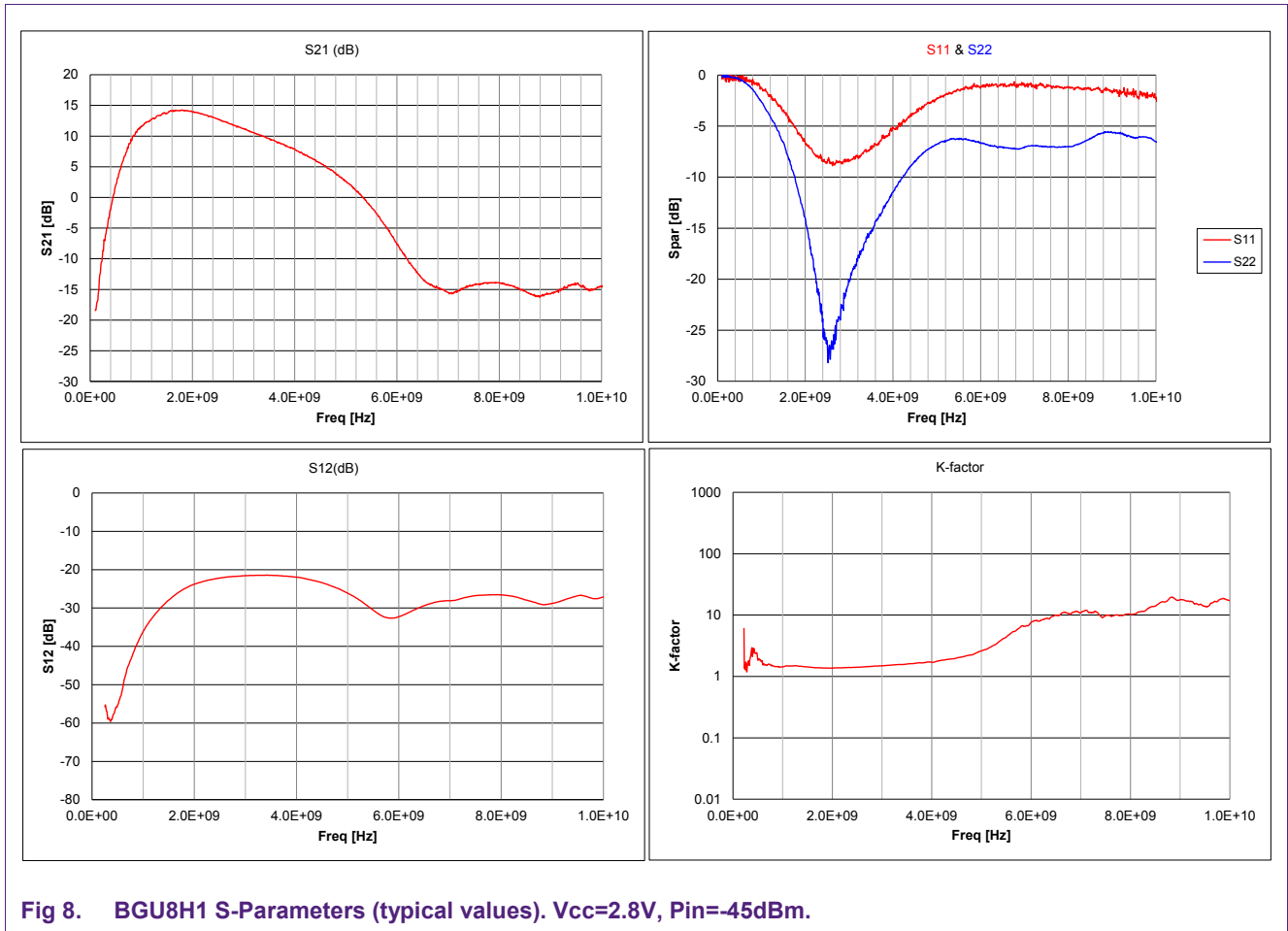


Fig 7. Setup diagram for 50Ω LNA-board NF-Measurements.

7. Evaluation Board Tests

7.1 S-Parameters

The measured S-Parameters and stability factor K are given in the figures below. For the measurements, a BGU8H1-LNA EVB is used ((see Fig 5). Measurements have been carried out using the setup shown in Fig 6.



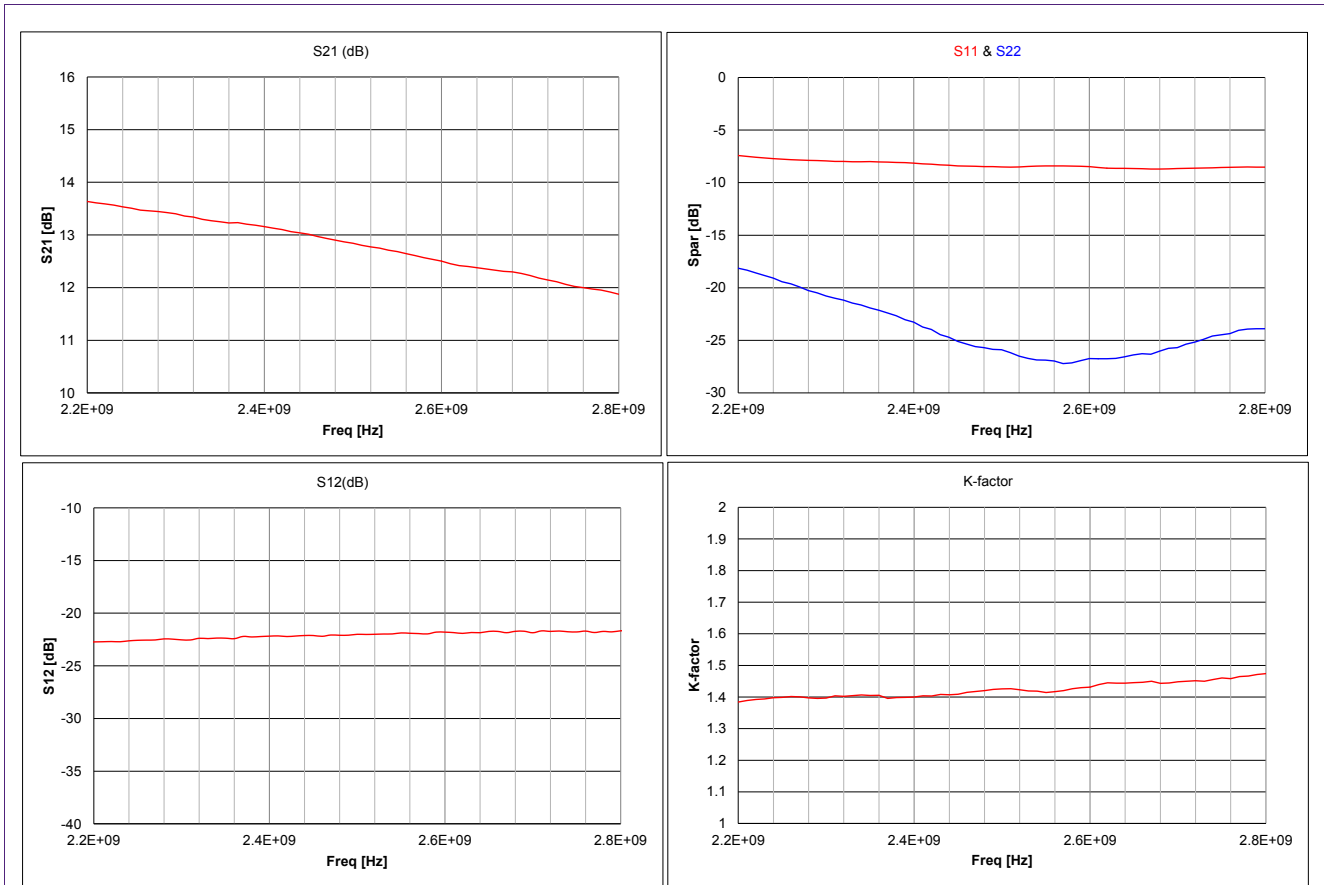


Fig 9. BGU8H1 S-Parameters (typical values). Vcc=2.8V, Pin=-45dBm (freq. range zoomed in).

7.2 1dB gain compression

Strong in-band cell phone TX jammers can cause linearity problems and result in third-order intermodulation products in the LTE frequency band. In this chapter the effects of these strong signals is shown. For the measurements, a BGU8H1-LNA EVB is used ((see Fig 5). Measurements have been carried out using the setup shown in Fig 6

The gain as function of input power of the DUT was measured between port RFin and RFout of the EVB at the LTE center frequencies.

The figures below show the gain compression curves at LNA-board.

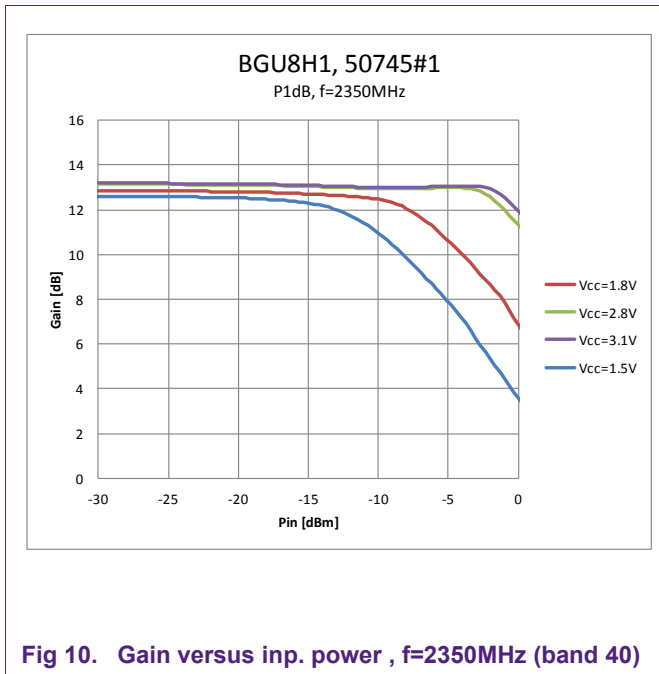


Fig 10. Gain versus inp. power , f=2350MHz (band 40)

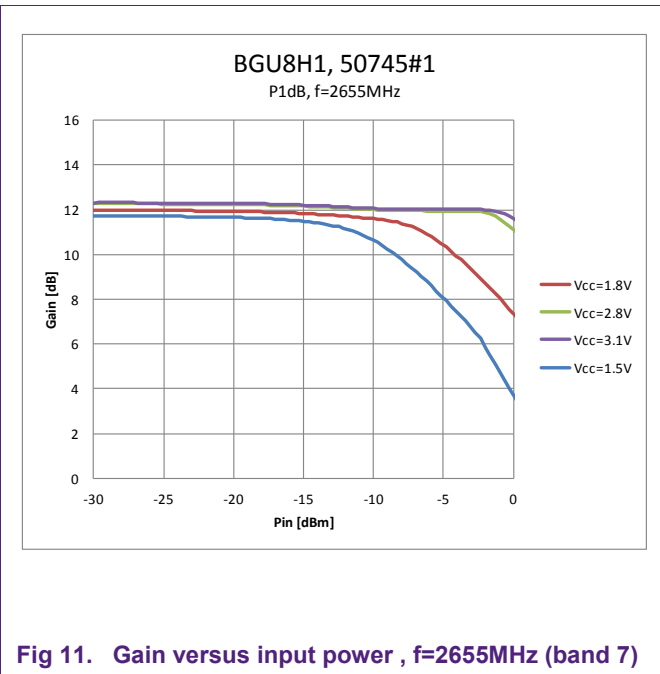


Fig 11. Gain versus input power , f=2655MHz (band 7)

7.3 2-Tone Test

The figures below show the spectra of the DUT caused by a 2-Tone input signal around the centre of the LTE-bands. For the measurements, a BGU8H1-LNA EVB is used ((see Fig 5). Measurements have been carried out using the setup shown in Fig 6.

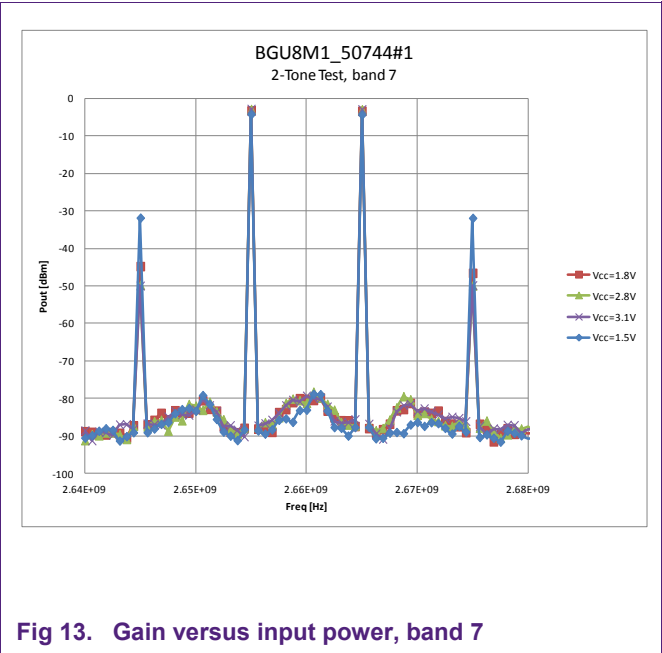
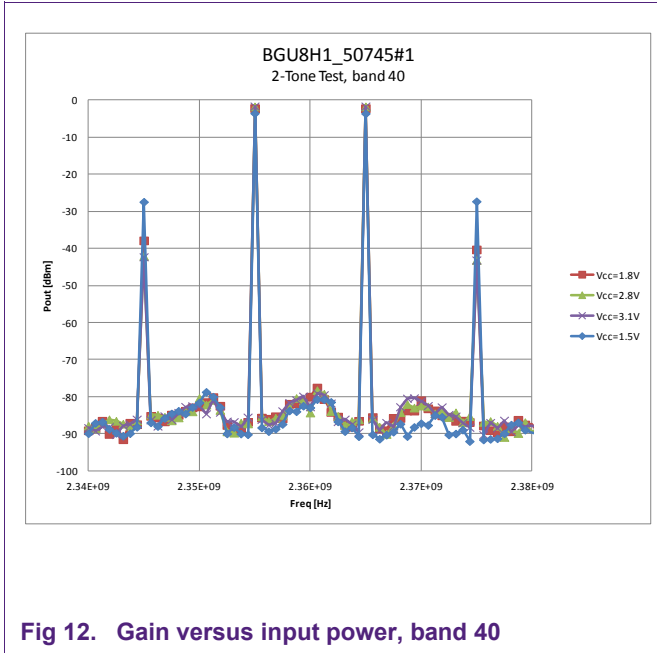


Fig 12. Gain versus input power, band 40

Fig 13. Gain versus input power, band 7

7.4 Enable Timing Test

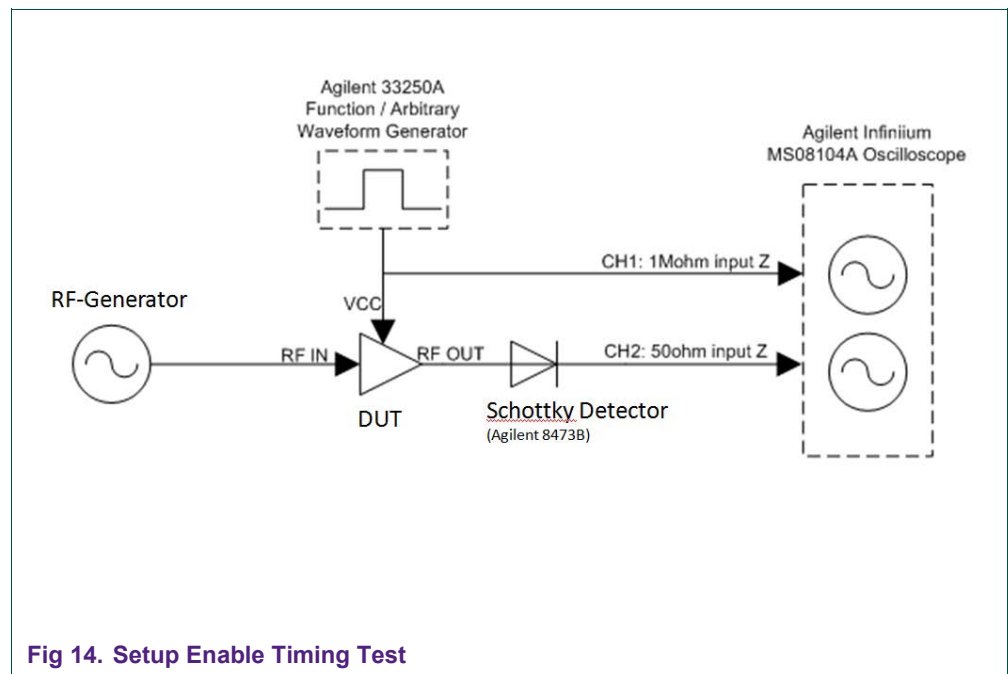
The following diagram shows the setup to test LNA Turn ON and Turn OFF time.

Set the waveform generator to square mode and the output amplitude at 3Vrms with high output impedance. The waveform generator has adequate output current to drive the LNA therefore no extra DC power supply is required which simplifies the test setup.

Set the RF signal generator output level to -20dBm between 2300 MHz and 2690 MHz and increase its level until the output DC on the oscilloscope is at 5mV on 1mV/division, the signal generator RF output level is approximately -3dBm.

It is very important to keep the cables as short as possible at input and output of the LNA so the propagation delay difference on cables between the two channels is minimized.

It is also critical to set the oscilloscope input impedance to 50ohm on channel 2 so the diode detector can discharge quickly to avoid a false result on the Turn OFF time testing.



The series capacitor will influence the Ton/Toff switching time. When the default value C2=1nF is used, Ton will approximately be 9us. By reducing C2 to 100pF, Ton is reduced to approximately 4µs (see Fig 15 and Fig 16).

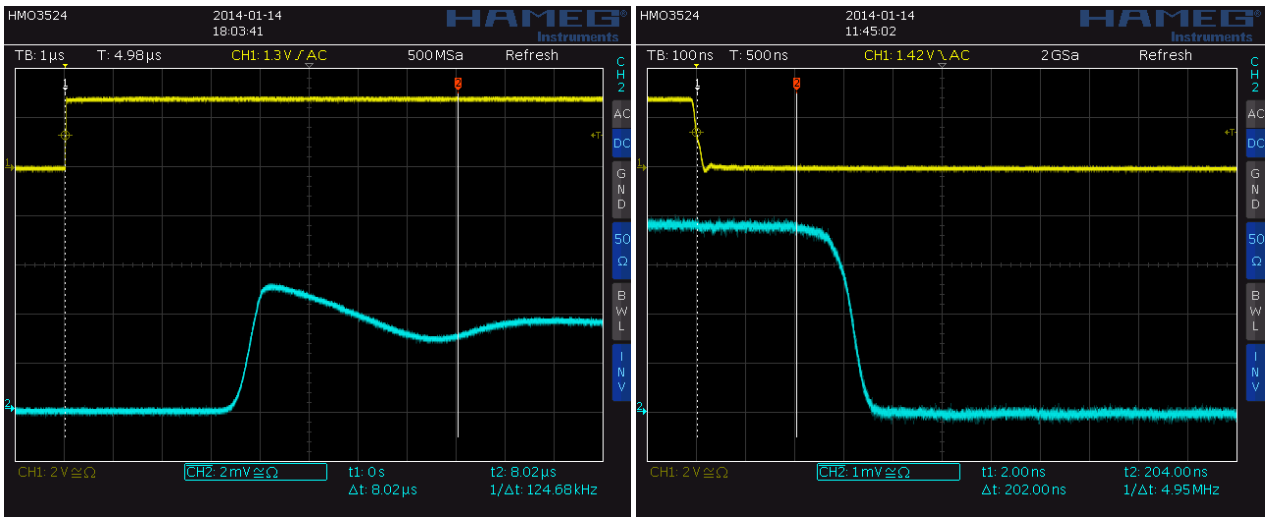


Fig 15. Results Enable Timing Test. Series capacitor C2=1nF. Ton~9µs (left) and Toff~200ns (right).

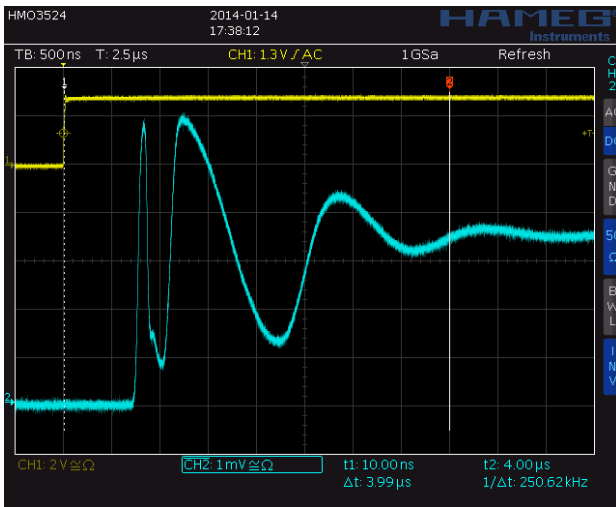


Fig 16. Results Enable Timing Test. Series capacitor C2=100pF. Ton~4µs (left).

8. Typical LNA evaluation board results

Table 2. Typical results measured on the evaluation Board.

| Typical LNA evaluation board results Temp = 25 °C | | | | | | | | |
|--|-------------|---------------|-------|------|------|------|------|------|
| Parameter | Freq. [MHz] | Symbol | | | | | Unit | Note |
| Supply Voltage | | Vcc | 1.5 | 1.8 | 2.8 | 3.1 | V | |
| Supply Current | | Icc | 4.3 | 4.8 | 5.0 | 5.0 | mA | |
| Noise Figure | 2350 | NF | 1.0 | 1.0 | 1.0 | 1.0 | dB | [1] |
| | 2655 | | 1.2 | 1.2 | 1.1 | 1.1 | | |
| Power Gain | 2350 | Gp | 12.7 | 13.0 | 13.0 | 13.3 | dB | |
| | 2655 | | 11.7 | 12.0 | 12.5 | 12.3 | | |
| Input Return Loss | 2350 | RLin | 7 | 8 | 9 | 8 | dB | |
| | 2655 | | 8 | 8 | 9 | 9 | | |
| Output Return Loss | 2350 | RLout | 22 | 20 | 20 | 23 | dB | |
| | 2655 | | 27 | 20 | 20 | 26 | | |
| Reverse Isolation | 2350 | ISOrev | 22 | 20 | 22 | 22 | dB | |
| | 2655 | | 22 | 20 | 22 | 22 | | |
| Input 1dB Gain Compression | 2350 | Pi1dB | -11.7 | -8.0 | -3.0 | -0.5 | dBm | |
| | 2655 | | -10.2 | -7.0 | -1.0 | 0.6 | | |
| Output 1dB Gain Compression | 2350 | Po1dB | 0.0 | 4.0 | 9.0 | 11.8 | dBm | |
| | 2655 | | 0.5 | 4.0 | 10.5 | 11.9 | | |
| Input third order intercept point | 2350 | IIP3 | -2.7 | 2.0 | 6.0 | 5.4 | dBm | [2] |
| | 2655 | | -0.8 | 5.0 | 8.0 | 8.6 | | |
| Output third order intercept point | 2350 | OIP3 | 10.0 | 15.0 | 19.0 | 18.7 | dBm | [2] |
| | 2655 | | 10.9 | 17.0 | 20.5 | 20.9 | | |
| Power settling time | | Ton | 4 | 4 | 4 | 4 | µs | |
| | | Toff | 1 | 1 | 1 | 1 | µs | |

[1] Including PCB losses
 [2] f = f_center_band; Delta_f=10MHz
 Pin_f1 = Pin_f2 = -15 dBm

9. Improved in band blocking performance modification

In some cases a strong in-band jamming signal is present, reducing the sensitivity. This in band blocking test case is illustrated below in Fig 17. A jamming signal causes an increase of the noise-floor closely around the jamming frequency, which reduces the sensitivity for a wanted signal overlapping with the noise band.

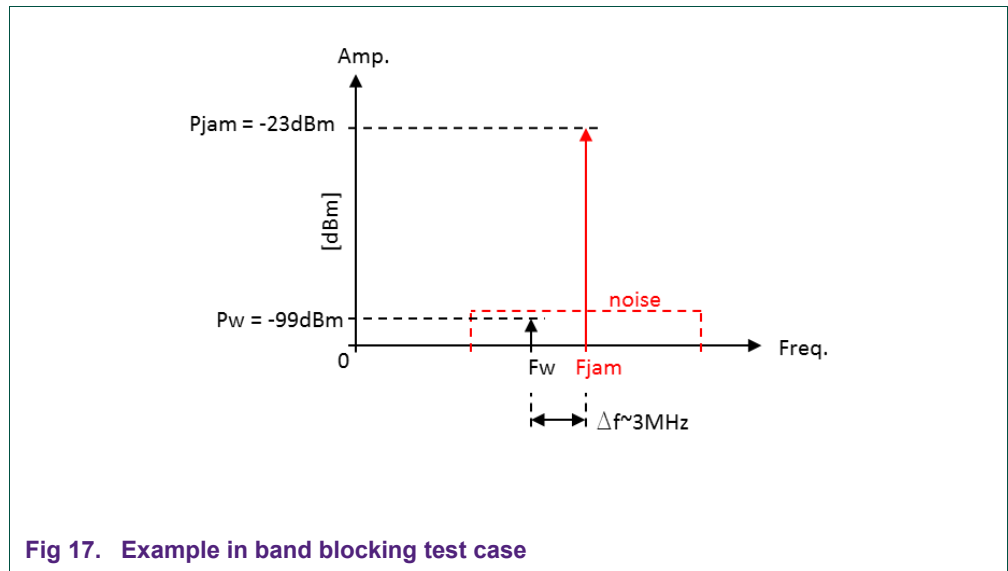


Fig 17. Example in band blocking test case

A solution is to make a low impedance path for low frequencies at the input of the LNA. This can be done by an additional shunt inductor $L2$ with a high value, as shown in the circuit of Fig 18 and board detail in Fig 19 ($L1$ and $C2$ have been swapped compared with Fig 2 to avoid a DC-path between R_{Fin} and GND). For $L2$ a Murata LQW15 wire wound inductor with a value of 82nH is used.

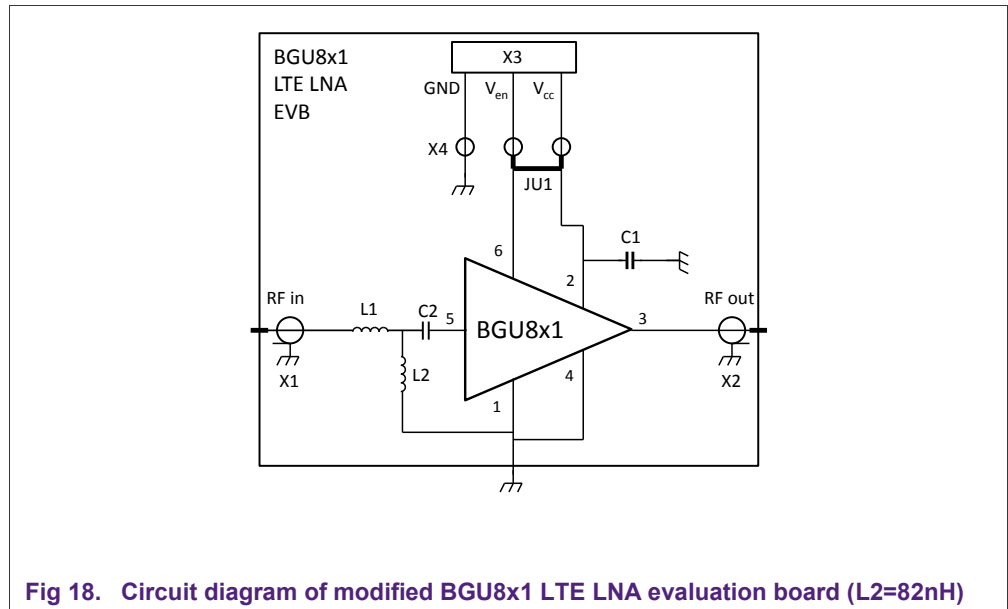


Fig 18. Circuit diagram of modified BGU8x1 LTE LNA evaluation board ($L2=82\text{nH}$)

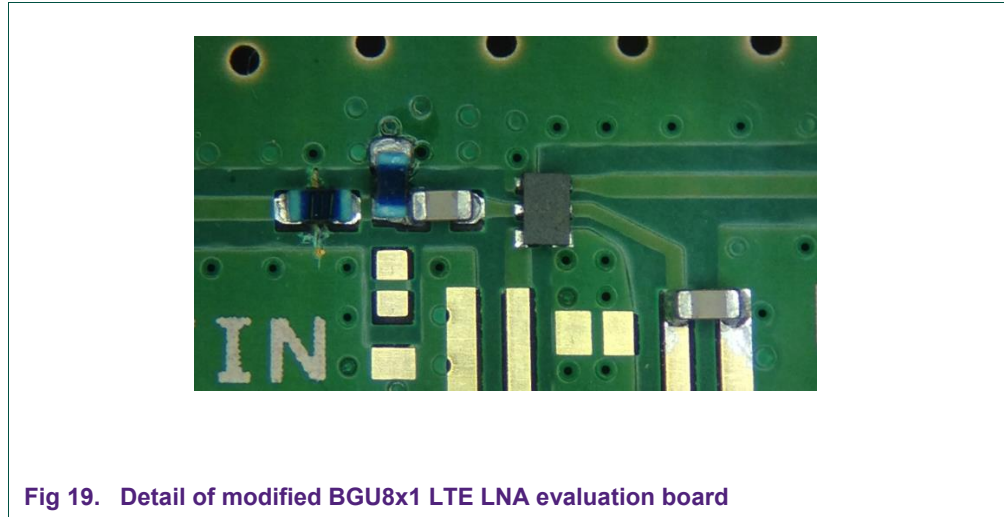


Fig 19. Detail of modified BGU8x1 LTE LNA evaluation board

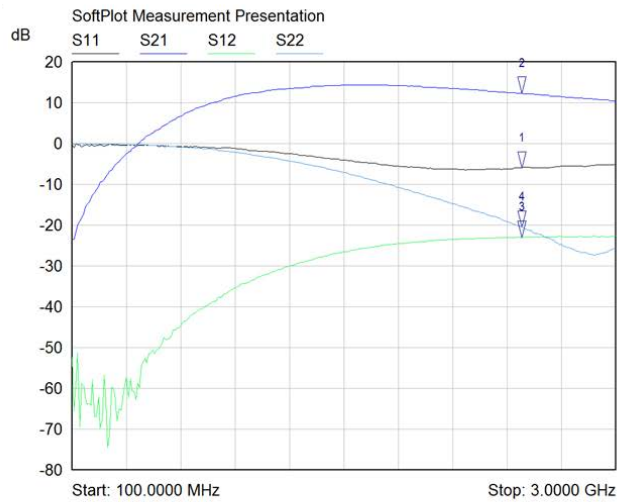
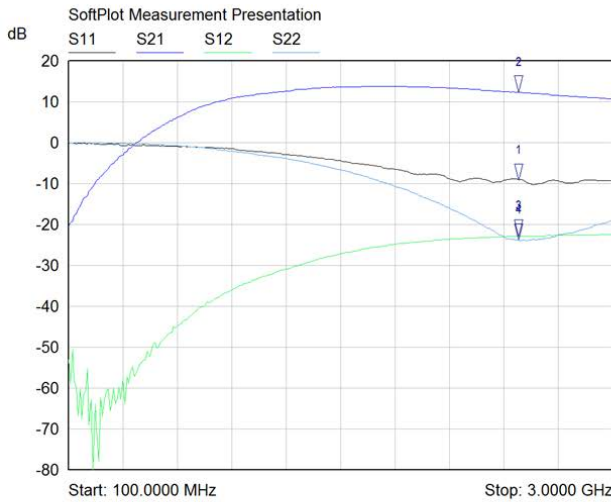
The measured performance is given in Fig 20 and Table 3. The Gain and the NF are almost equal for both versions.

EVB 95216#50

Vcc=2.8V, Default match.

EVB 157716#1

Vcc=2.8V, Modified input.



| Mkr | Trace | X-Axis | Value | Notes |
|-----|-------|------------|-----------|-------|
| 1 | S11 | 2.5000 GHz | -9.10 dB | |
| 2 | S21 | 2.5000 GHz | 12.32 dB | |
| 3 | S12 | 2.5000 GHz | -22.81 dB | |
| 4 | S22 | 2.5000 GHz | -23.91 dB | |

| Mkr | Trace | X-Axis | Value | Notes |
|-----|-------|------------|-----------|-------|
| 1 | S11 | 2.5000 GHz | -6.00 dB | |
| 2 | S21 | 2.5000 GHz | 12.26 dB | |
| 3 | S12 | 2.5000 GHz | -22.94 dB | |
| 4 | S22 | 2.5000 GHz | -20.43 dB | |

Fig 20. Measured performance modified BGU8x1 LTE LNA evaluation board

Table 3. Typical results measured on the modified evaluation Board.

| | | Typical LNA evaluation board results | | | | | | |
|--|-------------|--------------------------------------|-------------|-------------|-----------|------|-------|--|
| Temp [°C] | 25 | BU8H1_157716 | | | | | | |
| P_Spar [dBm] | -45 | 3 EVB's | | | | | | |
| Parameter | | Symbol | Default typ | L2=82nH typ | Delta typ | Unit | Notes | |
| | Freq. [MHz] | | | | | | | |
| Supply Voltage | | Vcc | 2.80V | 2.80V | 2.80V | V | | |
| Noise Figure | 2350 | NF | | | | dB | | |
| | 2500 | | 1.05 | 1.04 | -0.01 | | | |
| | 2655 | | | | | | | |
| Power Gain | 2350 | Gp | 13.0 | 12.8 | -0.1 | dB | | |
| | 2500 | | 12.5 | 12.3 | -0.2 | | | |
| | 2655 | | 12.0 | 11.7 | -0.2 | | | |
| Input Return Loss | 2350 | RLin | 8.7 | 6.2 | -2.5 | dB | | |
| | 2500 | | 8.3 | 6.0 | -2.3 | | | |
| | 2655 | | 8.5 | 5.9 | -2.6 | | | |
| Output Return Loss | 2350 | RLout | 20.6 | 17.9 | -2.7 | dB | | |
| | 2500 | | 22.7 | 20.4 | -2.2 | | | |
| | 2655 | | 23.5 | 23.6 | 0.1 | | | |
| Reverse Isolation | 2350 | ISOrev | 22.7 | 22.9 | 0.2 | dB | | |
| | 2500 | | 22.5 | 22.7 | 0.2 | | | |
| | 2655 | | 22.4 | 22.6 | 0.2 | | | |
| Input 1dB Gain Compression | 2350 | Pi1dB | -0.8 | -0.3 | 0.5 | dBm | | |
| | 2500 | | -0.3 | 0.5 | 0.8 | | | |
| | 2655 | | 0.4 | 1.0 | 0.6 | | | |
| Output 1dB Gain Compression | 2350 | Po1dB | 11.1 | 11.5 | 0.4 | dBm | | |
| | 2500 | | 11.2 | 11.8 | 0.7 | | | |
| | 2655 | | 11.3 | 11.7 | 0.4 | | | |
| Input third order intercept point (average Isb&usb) | 2350 | IIP3 | 4.2 | 4.3 | 0.1 | dBm | | |
| | 2500 | | 5.7 | 6.3 | 0.6 | | | |
| | 2655 | | 5.7 | 6.2 | 0.5 | | | |
| Output third order intercept point (average Isb&usb) | 2350 | OIP3 | 17.4 | 17.6 | 0.2 | dBm | | |
| | 2500 | | 18.1 | 18.8 | 0.7 | | | |
| | 2655 | | 17.4 | 18.1 | 0.7 | | | |

Note: Noise Figure is including PCB losses.

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11. List of figures

| | | |
|---------|---|----|
| Fig 1. | BGU8x1 LTE LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1)..... | 3 |
| Fig 2. | Circuit diagram of the BGU8x1 LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1) | 4 |
| Fig 3. | Printed-Circuit Board layout of the BGU8x1LNA evaluation board (used for BGU8L1, BGU8M1 and BGU8H1) | 5 |
| Fig 4. | Stack of the PCB material | 5 |
| Fig 5. | Evaluation board including its connections | 8 |
| Fig 6. | 2-Tone Setup for 50Ω LNA board tests (S-Parameters, P1dB and 2-Tone-tests) | 8 |
| Fig 7. | Setup diagram for 50Ω LNA-board NF-Measurements. | 9 |
| Fig 8. | BGU8H1 S-Parameters (typical values). Vcc=2.8V, Pin=-45dBm..... | 10 |
| Fig 9. | BGU8H1 S-Parameters (typical values). Vcc=2.8V, Pin=-45dBm (freq. range zoomed in). | 11 |
| Fig 10. | Gain versus inp. power , f=2350MHz (band 40) | 12 |
| Fig 11. | Gain versus input power , f=2655MHz (band 7) | 12 |
| Fig 12. | Gain versus input power, band 40 | 13 |
| Fig 13. | Gain versus input power, band 7 | 13 |
| Fig 14. | Setup Enable Timing Test..... | 14 |
| Fig 15. | Results Enable Timing Test. Series capacitor C2=1nF. Ton~9μs (left) and Toff~200ns (right). | 15 |
| Fig 16. | Results Enable Timing Test. Series capacitor C2=100pF. Ton~4μs (left). | 15 |
| Fig 17. | Example in band blocking test case | 17 |
| Fig 18. | Circuit diagram of modified BGU8x1 LTE LNA evaluation board (L2=82nH) | 17 |
| Fig 19. | Detail of modified BGU8x1 LTE LNA evaluation board..... | 18 |
| Fig 20. | Measured performance modified BGU8x1 LTE LNA evaluation board | 18 |

12. List of tables

Table 1. BOM of the BGU8H1 LTE LNA evaluation board6

Table 2. Typical results measured on the evaluation Board. 16

Table 3. Typical results measured on the modified evaluation Board. 19

13. Contents

| | | |
|------------|---|-----------|
| 1. | Introduction | 3 |
| 2. | General description..... | 4 |
| 3. | BGU8H1 LTE LNA evaluation board | 4 |
| 3.1 | Application Circuit | 4 |
| 3.2 | PCB Layout..... | 5 |
| 4. | Bill of materials..... | 6 |
| 4.1 | BGU8H1..... | 6 |
| 4.2 | Series inductor | 6 |
| 5. | Required Equipment | 7 |
| 6. | Connections and setup..... | 7 |
| 7. | Evaluation Board Tests | 10 |
| 7.1 | S-Parameters | 10 |
| 7.2 | 1dB gain compression..... | 12 |
| 7.3 | 2-Tone Test..... | 13 |
| 7.4 | Enable Timing Test | 14 |
| 8. | Typical LNA evaluation board results | 16 |
| 9. | Improved in band blocking performance modification | 17 |
| 10. | Legal information | 20 |
| 10.1 | Definitions | 20 |
| 10.2 | Disclaimers..... | 20 |
| 10.3 | Trademarks..... | 20 |
| 11. | List of figures..... | 21 |
| 12. | List of tables | 22 |
| 13. | Contents..... | 23 |

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